

COCO 2020  
July 16, 2020



## Bounds on the List Size of Successive Cancellation List Decoding

Mustafa Cemil Coşkun<sup>1,2</sup>

Joint work with H. D. Pfister<sup>3</sup>

<sup>1</sup>German Aerospace Center

<sup>2</sup>Technical University of Munich

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A photograph of the Earth from space, showing the curvature of the planet, blue oceans, green landmasses, and white clouds. The text 'Knowledge for Tomorrow' is overlaid on the right side of the image.

Knowledge for Tomorrow

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## **Average** List Size of Successive Cancellation **Inactivation** Decoding

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Knowledge for Tomorrow

# Outline

- Introduction
- Preliminaries
- The Binary Erasure Channel
- Numerical Results
- Conclusions



# Polar Codes

IEEE TRANSACTIONS ON INFORMATION THEORY, VOL. 55, NO. 7, JULY 2009

3051

## Channel Polarization: A Method for Constructing Capacity-Achieving Codes for Symmetric Binary-Input Memoryless Channels

Erdal Arkan, *Senior Member, IEEE*

**Abstract**—A method is proposed, called channel polarization, to construct code sequences that achieve the symmetric capacity  $I(W)$  of any given binary-input discrete memoryless channel (B-DMC)  $W$ . The symmetric capacity is the highest rate achievable subject to using the input letters of the channel with equal probability. Channel polarization refers to the fact that it is pos-

### A. Preliminaries

We write  $W : \mathcal{X} \rightarrow \mathcal{Y}$  to denote a generic B-DMC with input alphabet  $\mathcal{X}$ , output alphabet  $\mathcal{Y}$ , and transition probabilities  $W(y|x)$ ,  $x \in \mathcal{X}$ ,  $y \in \mathcal{Y}$ . The input alphabet  $\mathcal{X}$  will always be  $\{0, 1\}$ , the output alphabet and the transition probabilities may

- Capacity-achieving on binary memoryless symmetric (BMS) channels with low encoding/decoding complexity<sup>1</sup>

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<sup>1</sup>E. Arkan, "Channel polarization: A method for constructing capacity-achieving codes for symmetric binary-input memoryless channels," *T-IT*, Jul. 2009.



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- Capacity-achieving on binary memoryless symmetric (BMS) channels with low encoding/decoding complexity<sup>1</sup>
- But successive cancellation (SC) decoding performs poorly for small blocks

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# Successive List Cancellation Decoding

IEEE TRANSACTIONS ON INFORMATION THEORY, VOL. 61, NO. 5, MAY 2015

2213

## List Decoding of Polar Codes

Ido Tal, *Member, IEEE* and Alexander Vardy, *Fellow, IEEE*

**Abstract**—We describe a successive-cancellation list decoder for polar codes, which is a generalization of the classic successive-cancellation decoder of Arıkan. In the proposed list decoder,  $L$  decoding paths are considered concurrently at each decoding stage, where  $L$  is an integer parameter. At the end of the decoding process, the most likely among the  $L$  paths is selected as the single codeword at the decoder output. Simulations show that the resulting performance is very close to that of maximum-likelihood decoding, even for moderate values of  $L$ . Alternatively, if a genie is allowed to pick the transmitted codeword from the list, the results are comparable with the performance of current state-of-the-art LDPC codes. We show that such a genie can be easily implemented using simple CRC precoding. The specific list-decoding algorithm that achieves this performance doubles the number of decoding paths for each information bit, and then uses a pruning procedure to discard all but the  $L$  most likely paths. However, straightforward implementation of this

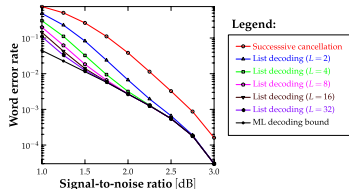


Fig. 1. List-decoding performance for a polar code of length  $n = 2048$  and rate  $R = 0.5$  on the BPSK-modulated Gaussian channel. The code was constructed using the methods of [15], with optimization for  $E_b/N_0 = 2$  dB.

- SC list (SCL) decoding with CRC and large list-size performs very well and matches maximum-likelihood (ML)<sup>2</sup>

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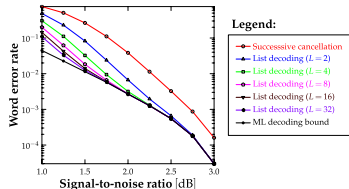


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- SC list (SCL) decoding with CRC and large list-size performs very well and matches maximum-likelihood (ML)<sup>2</sup>
- Can also be used to decode other codes (e.g., Reed–Muller codes)

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## Motivating Question

- What list size is **sufficient to approach ML decoding** performance for a given polar code and channel?





## Motivating Question

- What list size is **sufficient to approach ML decoding** performance for a given polar code and channel?
  - Can be attacked via simulation but **quite complex for long codes and lists**
  - But, simulation **unlikely to provide insight** into the question
  - A theoretical answer **might enable better code designs** for SCL decoding



## Summary of Results

- In this talk, we focus on the binary erasure channel (BEC)

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  - The random dimension sequence of this subspace can be approximated by a Markov chain
  - For a fixed number of erasures, the **approximation is reasonably accurate**

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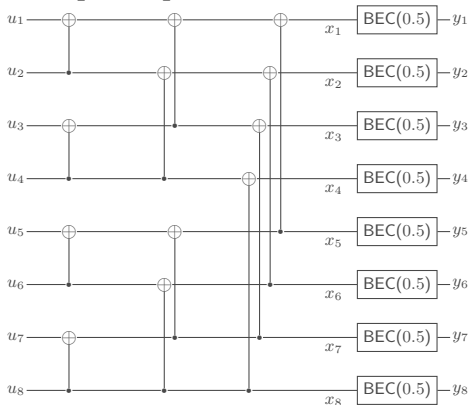
## Polar Codes and Density Evolution

$$x_1^n = u_1^n \mathbf{G}_2^{\otimes m} \quad \text{where} \quad \mathbf{G}_2 \triangleq \begin{bmatrix} 1 & 0 \\ 1 & 1 \end{bmatrix} \quad \text{and} \quad n = 2^m$$



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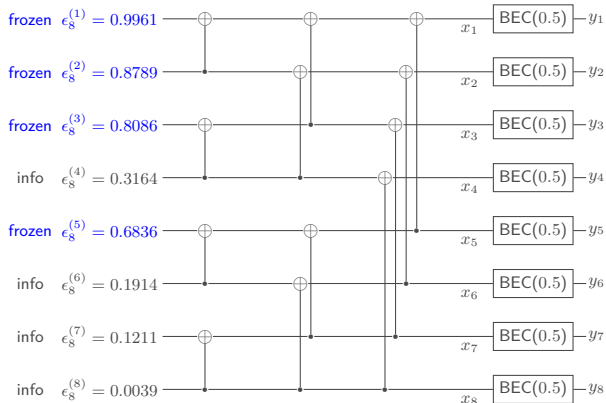
Effective Channel  $i$

$$W_N^{(i)}(y_1^n, u_1^{i-1} | u_i)$$

$$H(W_N^{(i)}) \triangleq H(U_i | Y_1^N, U_1^{i-1}) \\ = \epsilon_N^{(i)}$$

$$\mathcal{F} = \{1, 2, 3, 5\}$$

$$\mathcal{A} = \{4, 6, 7, 8\}$$





## Dynamic Frozen Bits

- The value of a frozen bit can also be set to a **linear combination of previous information bits** (rather than a fixed 0 or 1 value)<sup>4</sup>

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- A frozen bit whose value depends on past inputs is called **dynamic**
- SC/SCL decoding easily modified for polar codes with dynamic frozen bits
- **Any binary linear block code** can be represented as a polar code with dynamic frozen bits!

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  - If SC decoding step outputs erasure, **inactivate** the bit and add basis vector
  - Later messages in decoder are functions of inactivated bits (i.e., basis vectors)
  - If SC decoding of frozen bit is an unerased message, then resulting equation may allow one to **consolidate** the basis (i.e., remove a basis vector)



## The Uncertainty Dimension

- For a fixed  $y_1^N$ , the subspace dimension is  $d_m(y_1^N)$

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$$d_m(y_1^N) = H\left(U_{\mathcal{A}^{(m)}} \mid Y_1^N = y_1^N, U_{\mathcal{F}^{(m)}}\right) \text{ where } \mathcal{A}^{(m)} \triangleq \mathcal{A} \cap [m] \text{ and } \mathcal{F}^{(m)} \triangleq \mathcal{F} \cap [m]$$



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Next goal is to understand the evolution of the random sequence  $D_m$

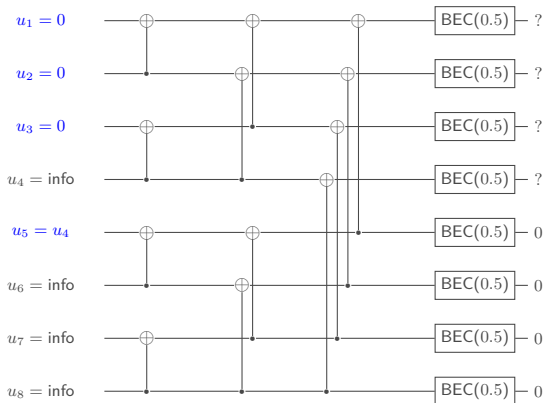
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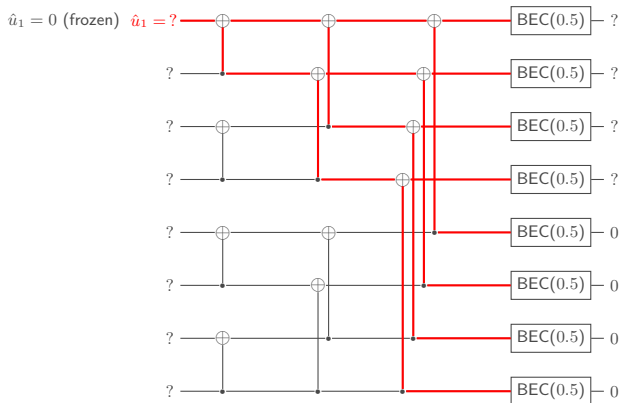
## Successive Cancellation Inactivation Decoding

Example:  $u_1 = u_2 = u_3 = 0$ ,  $u_5 = u_4$  (frozen bits)



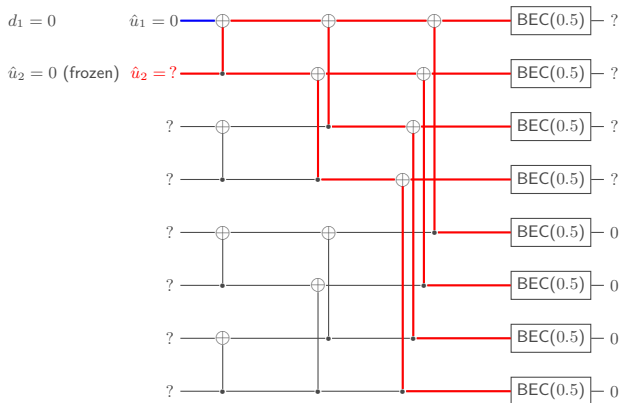
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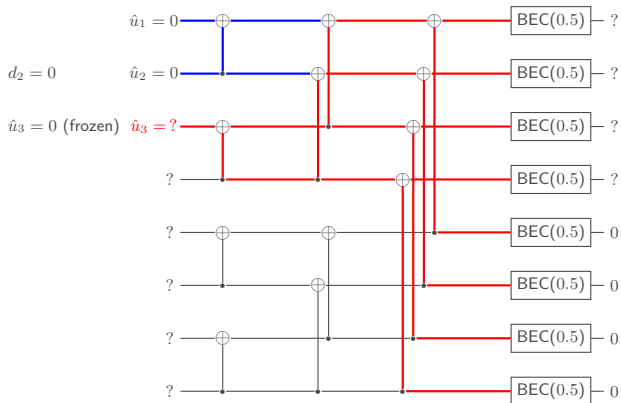
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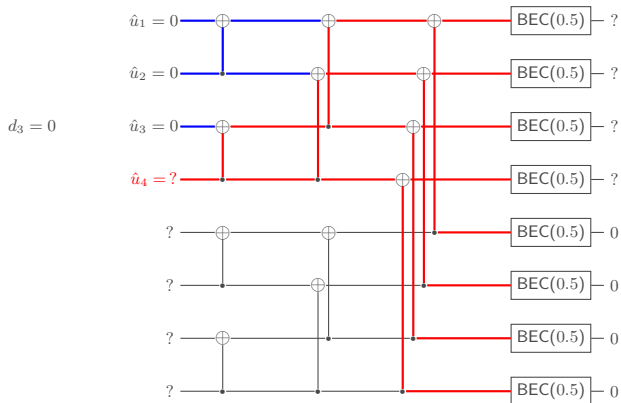
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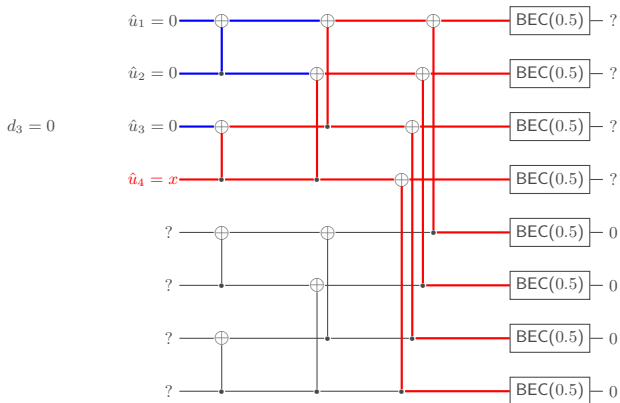
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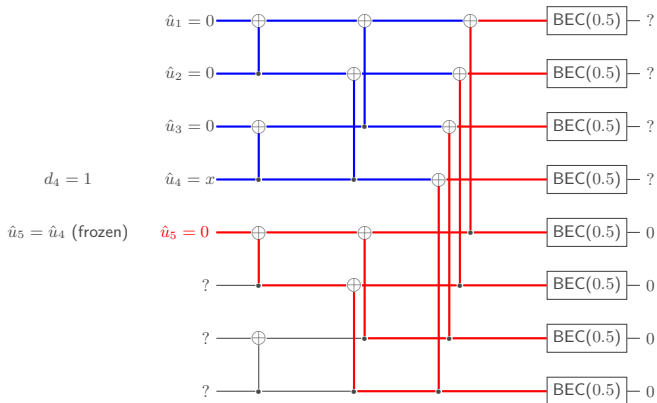
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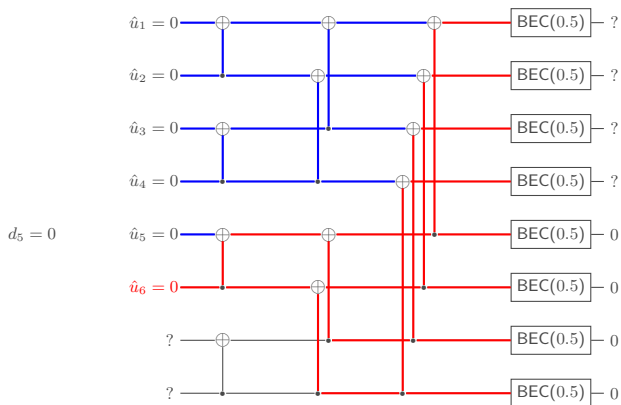
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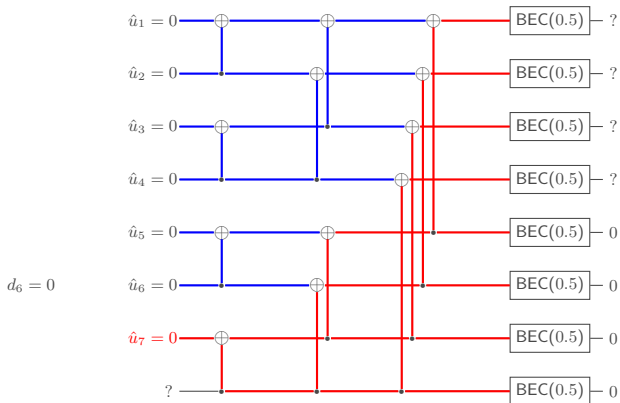
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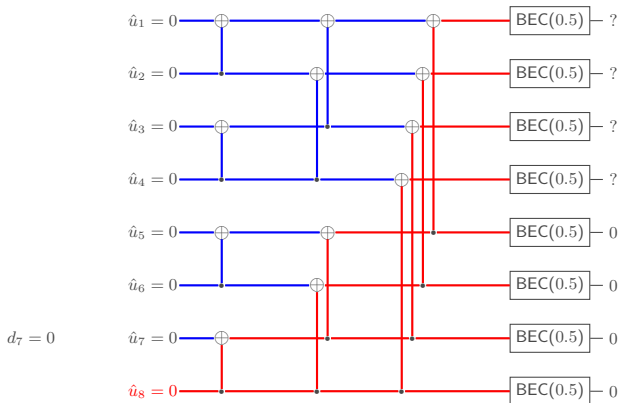
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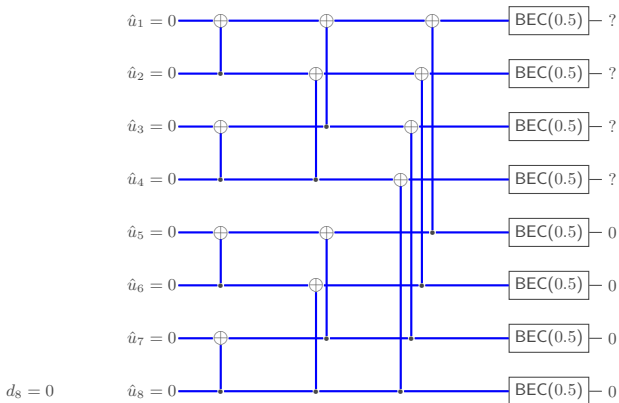
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## Evolution of the Uncertainty Dimension

- If  $U_m$  is an information bit, then
  - ✓ If decoder outputs an erasure, then  $d_m(y_1^N) = d_{m-1}(y_1^N) + 1$
  - ✓ Else, it outputs affine function and  $d_m(y_1^N) = d_{m-1}(y_1^N)$



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  - ✓ Else, it outputs affine function:
    - × If **consolidation**:  
 $d_m(y_1^N) = d_{m-1}(y_1^N) - 1$
    - × Else, **no consolidation**:  
 $d_m(y_1^N) = d_{m-1}(y_1^N)$



## The Markov Chain Approximation

- The random sequence  $D_1, \dots, D_N$  can be approximated by an inhomogeneous Markov chain with transition probabilities

$P_{i,j}^{(m)} \approx \Pr(D_m = j \mid D_{m-1} = i)$  where

$$P_{i,j}^{(m)} = \begin{cases} \epsilon_N^{(m)} & \text{if } m \in \mathcal{A}, j = i + 1 \\ 1 - \epsilon_N^{(m)} & \text{if } m \in \mathcal{A}, j = i \\ \epsilon_N^{(m)} + (1 - \epsilon_N^{(m)}) 2^{-D_{m-1}} & \text{if } m \in \mathcal{F}, j = i \\ (1 - \epsilon_N^{(m)}) (1 - 2^{-D_{m-1}}) & \text{if } m \in \mathcal{F}, j = i - 1 \end{cases}$$



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- $2^{-D}$  is probability a random  $D$ -variable equation has all zero coefficients



## d-RM and PAC Codes

- d-RM code ensemble<sup>3</sup>:
  - Let  $\mathcal{A}$  be the information indices of an RM code
  - $u_i$  is an information bit if  $i \in \mathcal{A}$
  - $u_i = \sum_{j \in \mathcal{A}(i)} A_{ij} u_j$  if  $i \in \mathcal{F}$ ,  
where  $A_{ij}$  iid  $\sim$  Bernoulli(0.5)

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<sup>5</sup>E. Arkan, "From sequential decoding to Channel Polarization and Back Again," Shannon Lecture, Jun. 2019.



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### ■ PAC code<sup>5</sup>:

- Given set  $\mathcal{A}$  and convolutional code with rate 1 and memory  $\nu$
- $u_i$  is an information bit if  $i \in \mathcal{A}$
- $u_i = g_i(u_{i-\nu}^{i-1})$  if  $i \in \mathcal{F}$  where  $g_i(\cdot)$  defined by CC

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<sup>5</sup>E. Arkan, "From sequential decoding to Channel Polarization and Back Again," *Shannon Lecture*, Jun. 2019.





## d-RM and PAC Codes

### ■ d-RM code ensemble<sup>3</sup>:

- Let  $\mathcal{A}$  be the information indices of an RM code
- $u_i$  is an information bit if  $i \in \mathcal{A}$
- $u_i = \sum_{j \in \mathcal{A}(i)} A_{ij} u_j$  if  $i \in \mathcal{F}$ ,  
where  $A_{ij}$  iid  $\sim$  Bernoulli(0.5)

### ■ PAC code<sup>5</sup>:

- Given set  $\mathcal{A}$  and convolutional code with rate 1 and memory  $\nu$
- $u_i$  is an information bit if  $i \in \mathcal{A}$
- $u_i = g_i(u_{i-\nu}^{i-1})$  if  $i \in \mathcal{F}$  where  $g_i(\cdot)$  defined by CC

Choosing  $\mathcal{A}$  like an RM code for the PAC code (as Arıkan did) makes two codes differ only in the dynamic frozen bit constraints!

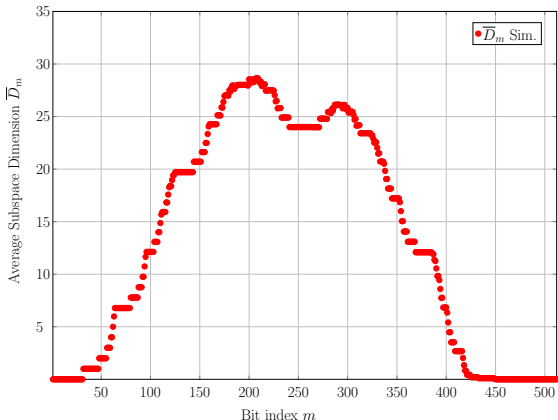
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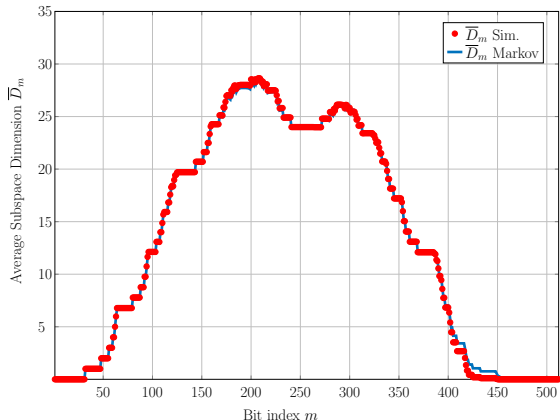
## (512, 256) d-RM Code

A fixed-weight BEC with exactly  $\text{round}(512 \times 0.48) = 246$  erasures



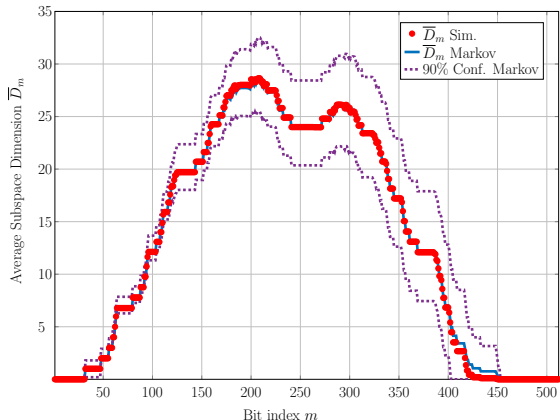
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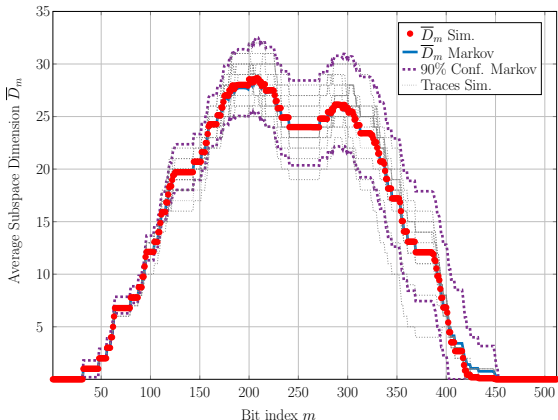
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## Summary

- “What list size is sufficient to approach ML decoding performance under an SCL decoder?”
  - ✓ Good approximation proposed for the BEC that can be computed efficiently

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- What is not covered in this talk?<sup>6</sup>
  - ✓ The quantity  $d_m$  (a conditional entropy) can be used as proxy for uncertainty in SCL decoding for general BMS channels

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- Outlook and Future Work
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  - ⚠ Apply this technique to design longer codes with good SCL performance

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Thank you! Questions?

## (512, 256) Codes - Performance

